

## AM/FM PLL Frequency Synthesizer

## Overview

The LC72130 and LC72130M are PLL frequency synthesizers for use in tuners in radio cassette recorders and other products.

## Applications

PLL frequency synthesizer

## Functions

- High-speed programmable dividers
- FMIN: 10 to 160 MHz $\qquad$ pulse swallower (built-in divide-by-two prescaler)
- AMIN:

2 to 40 MHz $\qquad$ .pulse swallower
0.5 to 10 MHz $\qquad$ direct division

- IF counter
- IFIN: 0.4 to 12 MHz $\qquad$ AM/FM IF counter
- Reference frequencies
- Twelve selectable frequencies
(4.5 or 7.2 MHz crystal)
$1,3,5,9,10,3.125,6.25,12.5,15,25,50$ and 100 kHz
- Phase comparator
- Dead zone control
- Unlock detection
- Deadlock clear circuit
- Built-in MOS transistor for implementing an active lowpass filter (two systems)
- Inputs and outputs
- Dedicated output ports: five pins
- Input or output ports: two pins
- Clock time base output available
- Serial data I/O
- Supports CCB format communication with the system controller.
- Operating ranges
- Supply voltage 4.5 to 5.5 V
- Operating temperature. $\qquad$ -40 to $+85^{\circ} \mathrm{C}$
- Packages
- DIP24S, MFP24S
- CCB is a trademark of SANYO ELECTRIC CO., LTD.
- CCB is SANYO's original bus format and all the bus addresses are controlled by SANYO.


## Package Dimensions

unit: mm
3067-DIP24S

unit: mm
3112-MFP24S


## Pin Assignment



A03481

Block Diagram


## Specifications

Absolute Maximum Ratings at $\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}$

| Parameter | Symbol | Pins | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Maximum supply voltage | $\mathrm{V}_{\text {DD }}$ max | $\mathrm{V}_{\mathrm{DD}}$ | -0.3 to +7.0 | V |
| Maximum input voltage | $\mathrm{V}_{\text {IN }} 1$ max | CE, CL, DI, AIN1, AIN2 | -0.3 to +7.0 | V |
|  | $\mathrm{V}_{\text {IN }} 2$ max | XIN, FMIN, AMIN, IFIN | -0.3 to $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
|  | $\mathrm{V}_{\text {IN }} 3$ max | $\overline{\mathrm{O} 1}, \overline{\mathrm{O} 2}$ | -0.3 to +15 | V |
| Maximum output voltage | $\mathrm{V}_{0} 1$ max | DO | -0.3 to +7.0 | V |
|  | $\mathrm{V}_{\mathrm{O}} 2$ max | XOUT, PD1, PD2 | -0.3 to $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
|  | $\mathrm{V}_{\mathrm{O}} 3$ max | $\overline{\mathrm{BO} 1}$ to $\overline{\mathrm{BO} 5}, \overline{\mathrm{IO} 1}, \overline{\mathrm{IO} 2}$, AOUT1, AOUT2 | -0.3 to +15 | V |
| Maximum output current | 101 max | $\overline{\mathrm{BO} 1}$ | 0 to 3.0 | mA |
|  | $\mathrm{l}_{0} 2$ max | DO, AOUT1, AOUT2 | 0 to 6.0 | mA |
|  | $\mathrm{l}_{0} 3$ max | $\overline{\mathrm{BO} 2}$ to $\overline{\mathrm{BO} 5}, \overline{\mathrm{IO} 1}, \overline{\mathrm{IO} 2}$ | 0 to 10.0 | mA |
| Allowable power dissipation | Pd max | $\mathrm{Ta} \leq 85^{\circ} \mathrm{C}$ | $\begin{aligned} & \text { DIP24S: } 350 \\ & \text { MFP24S: } 200 \end{aligned}$ | mW |
| Operating temperature | Topr |  | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  | -55 to +125 | ${ }^{\circ} \mathrm{C}$ |


| Parameter | Symbol | Pins | Conditions | min | typ | max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage | $V_{\text {DD }}$ | $V_{\text {DD }}$ |  | 4.5 |  | 5.5 | V |
| Input high level voltage | $\mathrm{V}_{\text {IH }}{ }^{1}$ | CE, CL, DI |  | $0.7 \mathrm{~V}_{\mathrm{DD}}$ |  | 6.5 | V |
|  | $\mathrm{V}_{\mathrm{HH}}{ }^{2}$ | $\overline{\mathrm{O} 1, \mathrm{IO}}$ |  | $0.7 \mathrm{~V}_{\mathrm{DD}}$ |  | 13 | V |
| Input low level voltage | $\mathrm{V}_{\mathrm{IL}}$ | $\mathrm{CE}, \mathrm{CL}, \mathrm{DI}, \overline{\mathrm{IO} 1}, \overline{\mathrm{IO} 2}$ |  | 0 |  | $0.3 \mathrm{~V}_{\mathrm{DD}}$ | V |
| Output voltage | $\mathrm{V}_{\mathrm{O}} 1$ | DO |  | 0 |  | 6.5 | V |
|  | $\mathrm{V}_{\mathrm{O}} 2$ | $\overline{\mathrm{BO} 1}$ to $\overline{\mathrm{BO5},} \overline{\mathrm{IO} 1}, \overline{\mathrm{IO} 2}$, AOUT1, AOUT2 |  | 0 |  | 13 | V |
| Input frequency | $\mathrm{f}_{\mathrm{IN}}{ }^{1}$ | XIN | $\mathrm{V}_{\text {IN }}{ }^{1}$ | 1 |  | 8 | MHz |
|  | $\mathrm{f}_{\mathrm{IN}}{ }^{2}$ | FMIN | $\mathrm{V}_{\text {IN }}{ }^{2}$ | 10 |  | 160 | MHz |
|  | $\mathrm{f}_{1 \mathrm{~N}^{3}}$ | AMIN | $\mathrm{V}_{\text {IN }} 3, \mathrm{SNS}=1$ | 2 |  | 40 | MHz |
|  | $\mathrm{f}_{\mathrm{IN}}{ }^{4}$ | AMIN | $\mathrm{V}_{\text {IN }} 4, \mathrm{SNS}=0$ | 0.5 |  | 10 | MHz |
|  | $\mathrm{f}_{\mathrm{IN}} 5$ | IFIN | $\mathrm{V}_{\text {IN }} 5$ | 0.4 |  | 12 | MHz |
| Input amplitude | $\mathrm{V}_{\text {IN }} 1$ | XIN | ${ }_{\text {fin }}{ }^{1}$ | 400 |  | 1500 | mVrms |
|  | $\mathrm{V}_{\text {IN }} \mathrm{V}^{2-1}$ | FMIN | $\mathrm{f}=10$ to 130 MHz | 40 |  | 1500 | mVrms |
|  | $\mathrm{V}_{\text {IN }} \mathrm{V}^{2-2}$ | FMIN | $\mathrm{f}=130$ to 160 MHz | 70 |  | 1500 | mVrms |
|  | $\mathrm{V}_{1 \mathrm{~N}}{ }^{3}$ | AMIN | $\mathrm{f}_{\text {IN }}{ }^{3}$, SNS $=1$ | 40 |  | 1500 | mVrms |
|  | $\mathrm{V}_{\text {IN }} 4$ | AMIN | $\mathrm{f}_{\mathrm{IN}}{ }^{4}$, SNS $=0$ | 40 |  | 1500 | mVrms |
|  | $\mathrm{V}_{\text {IN }} 5$ | IFIN | $\mathrm{f}_{\text {IN }} 5$, IFS $=1$ | 40 |  | 1500 | mVrms |
|  | $\mathrm{V}_{1 \mathrm{~N}^{6}}$ | IFIN | $\mathrm{f}_{\mathrm{IN}} 6, \mathrm{IFS}=0$ | 70 |  | 1500 | mVrms |
| Oscillation-guaranteed crystal resonator | Xtal | XIN, XOUT | * | 4.0 |  | 8.0 | MHz |

## Allowable Operating Ranges at $\mathrm{Ta}=-\mathbf{4 0}$ to $+85^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}$

Note: * Recommended crystal oscillator CI values:
$\mathrm{CI} \leq 120 \Omega$ (For a 4.5 MHz crystal)
$\mathrm{Cl} \leq 70 \Omega$ (For a 7.2 MHz crystal)
However, since the oscillator circuit characteristics depend on the printed circuit board and component values actually used, we recommend requesting a circuit evaluation from the manufacturer of the crystal used.
<Sample Ocsillator Circuit>
Crystal oscillator: HC-49/U (manufactured by Kinseki, Ltd.), CL = 12 pF
$\mathrm{C} 1=\mathrm{C} 2=15 \mathrm{pF}$
The circuit constants for the crystal oscillator circuit depend on the crystal used, the printed circuit board pattern, and other items. Therefore we recommend consulting with the manufacturer of the crystal for evaluation and reliability.


Electrical Characteristics at $\mathbf{T a}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}$

| Parameter | Symbol | Pins | Conditions | min | typ | max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Built-in feedback resistance | Rf1 | XIN |  |  | 1.0 |  | $\mathrm{M} \Omega$ |
|  | Rf2 | FMIN |  |  | 500 |  | k $\Omega$ |
|  | Rf3 | AMIN |  |  | 500 |  | k $\Omega$ |
|  | Rf4 | IFIN |  |  | 250 |  | k $\Omega$ |
| Built-in pull-down resistor | Rpd1 | FMIN |  |  | 200 |  | $\mathrm{k} \Omega$ |
|  | Rpd2 | AMIN |  |  | 200 |  | k $\Omega$ |
| Hysteresis | $\mathrm{V}_{\mathrm{HIS}}$ | $\mathrm{CE}, \mathrm{CL}, \mathrm{DI}, \overline{\mathrm{IO} 1}, \overline{\mathrm{IO} 2}$ |  |  | $0.1 \mathrm{~V}_{\mathrm{DD}}$ |  | V |
| Output high level voltage | $\mathrm{V}_{\mathrm{OH}}{ }^{1}$ | PD1, PD2 | $1 \mathrm{O}=-1 \mathrm{~mA}$ | $\mathrm{V}_{\mathrm{DD}}-1.0$ |  |  | V |
| Output low level voltage | $\mathrm{V}_{\mathrm{OL}}{ }^{1}$ | PD1, PD2 | $\mathrm{IO}=1 \mathrm{~mA}$ |  |  | 1.0 | V |
|  | $\mathrm{V}_{\mathrm{OL}}{ }^{2}$ | $\overline{\mathrm{BO} 1}$ | $1 \mathrm{O}=0.5 \mathrm{~mA}$ |  |  | 0.5 | V |
|  |  |  | $\mathrm{IO}=1 \mathrm{~mA}$ |  |  | 1.0 | V |
|  | $\mathrm{V}_{\mathrm{OL}}{ }^{3}$ | DO | $\mathrm{IO}=1 \mathrm{~mA}$ |  |  | 0.2 | V |
|  |  |  | $\mathrm{IO}=5 \mathrm{~mA}$ |  |  | 1.0 | V |
|  | $\mathrm{V}_{\text {OL }} 4$ | $\overline{\mathrm{BO} 2}$ to $\overline{\mathrm{BO} 5}, \overline{\mathrm{IO}}, \overline{\mathrm{IO} 2}$ | $\mathrm{IO}=1 \mathrm{~mA}$ |  |  | 0.2 | V |
|  |  |  | $\mathrm{IO}=5 \mathrm{~mA}$ |  |  | 1.0 | V |
|  |  |  | $\mathrm{IO}=8 \mathrm{~mA}$ |  |  | 1.6 | V |
|  | $\mathrm{V}_{\text {OL }} 5$ | AOUT1, AOUT2 | $\mathrm{IO}=1 \mathrm{~mA}, \mathrm{AlN}=1.3 \mathrm{~V}$ |  |  | 0.5 | V |
| Input high level current | $\mathrm{I}_{1{ }^{1}}$ | CE, CL, DI | $\mathrm{V}_{1}=6.5 \mathrm{~V}$ |  |  | 5.0 | V |
|  | $\mathrm{I}_{1 \mathrm{H}^{2}}$ | $\overline{\mathrm{O} 1}, \overline{\mathrm{IO} 2}$ | $\mathrm{V}_{1}=13 \mathrm{~V}$ |  |  | 5.0 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{1 \mathrm{H}^{3}}$ | XIN | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{DD}}$ | 2.0 |  | 11 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{1 \mathrm{H}^{4}}$ | FMIN, AMIN | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{DD}}$ | 4.0 |  | 22 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{\mathrm{IH} 5}$ | IFIN | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{DD}}$ | 8.0 |  | 44 | $\mu \mathrm{A}$ |
|  | $1_{1 H^{6}}$ | AIN1, AIN2 | $\mathrm{V}_{\mathrm{I}}=6.5 \mathrm{~V}$ |  |  | 200 | nA |
| Input low level current | $\mathrm{I}_{\text {IL }} 1$ | CE, CL, DI | $\mathrm{V}_{1}=0 \mathrm{~V}$ |  |  | 5.0 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{\text {IL }}{ }^{\text {2 }}$ | $\overline{\mathrm{O} 1,} \overline{\mathrm{IO} 2}$ | $\mathrm{V}_{1}=0 \mathrm{~V}$ |  |  | 5.0 | $\mu \mathrm{A}$ |
|  | ILL3 | XIN | $\mathrm{V}_{1}=0 \mathrm{~V}$ | 2.0 |  | 11 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{\text {IL }} 4$ | FMIN, AMIN | $\mathrm{V}_{1}=0 \mathrm{~V}$ | 4.0 |  | 22 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{\text {IL }} 5$ | IFIN | $\mathrm{V}_{1}=0 \mathrm{~V}$ | 8.0 |  | 44 | $\mu \mathrm{A}$ |
|  | ILL $^{6}$ | AIN1, AIN2 | $\mathrm{V}_{1}=0 \mathrm{~V}$ |  |  | 200 | nA |
| Output off leakage current | loff1 | $\overline{\mathrm{BO}}$ to $\overline{\mathrm{BO5}}, \mathrm{AOUT} 1$, AOUT2, $\overline{\mathrm{IO}}, \overline{\mathrm{IO} 2}$ | $\mathrm{V}_{\mathrm{O}}=13 \mathrm{~V}$ |  |  | 5.0 | $\mu \mathrm{A}$ |
|  | $\mathrm{l}_{\mathrm{OFF}}{ }^{2}$ | DO | $\mathrm{V}_{\mathrm{O}}=6.5 \mathrm{~V}$ |  |  | 5.0 | $\mu \mathrm{A}$ |
| High level three-state off leakage current | lofft | PD1, PD2, | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{DD}}$ |  | 0.01 | 200 | nA |
| Low level three-state off leakage current | IOFFL | PD1, PD2 | $\mathrm{V}_{\mathrm{O}}=0 \mathrm{~V}$ |  | 0.01 | 200 | nA |
| Input capacitance | $\mathrm{Cl}_{\text {IN }}$ | FMIN |  |  | 6 |  | pF |
| Current drain | $I_{\text {DD }} 1$ | $V_{D D}$ | $\begin{aligned} & \hline \mathrm{XtaI}=7.2 \mathrm{MHz}, \\ & \mathrm{f}_{\mathrm{IN}}{ }^{2}=130 \mathrm{MHz}, \\ & \mathrm{~V}_{\mathrm{IN}^{2}-1}=40 \mathrm{mVrms} \end{aligned}$ |  | 5 | 10 | mA |
|  | $\mathrm{IDD}^{2}$ | $V_{D D}$ | PLL block stopped (PLL INHIBIT), <br> Xtal oscillator operating (Xtal $=7.2 \mathrm{MHz}$ ) |  | 0.5 |  | mA |
|  | $\mathrm{IDD}^{3}$ | $V_{D D}$ | PLL block stopped Xtal oscillator stopped |  |  | 10 | $\mu \mathrm{A}$ |

## Pin Functions

| Symbol | Pin No. | Type | Functions | Circuit configuration |
| :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { XIN } \\ & \text { XOUT } \end{aligned}$ | $\begin{gathered} 1 \\ 24 \end{gathered}$ | X'tal OSC | - Crystal resonator connection (4.5/7.2 MHz) |  |
| FMIN | 15 | Local oscillator signal input | - Serial data input: FMIN is selected when DVS is set to 1. <br> - The input frequency range is from 10 to 160 MHz . <br> - The signal is passed through a built-in divide-by-two prescaler and then supplied to the swallow counter. <br> - Although the range of divisor settings is from 272 to 65,535 , the actual divisor is twice the setting since there is also a built-in divide-by-two prescaler. | A02599 |
| AMIN | 14 | Local oscillator signal input | - Serial data input: AMIN is selected when DVS is set to 0 . <br> - Serial data input: When SNS is set to 1 : <br> - The input frequency range is from 2 to 40 MHz . <br> - The signal is supplied directly to the swallow counter. <br> - The range of divisor settings is from 272 to 65,535 and the actual divisor will be the value set. <br> - Serial data input: When SNS is set to 0 : <br> - The input frequency range is from 0.5 to 10 MHz . <br> - The signal is supplied directly to a 12 -bit programmable divider. <br> - The range of divisor settings is from 4 to 4,095 and the actual divisor will be the value set. | A02599 |
| CE | 3 | Chip enable | - Must be set high when serial data is input to the LC72130 (DI), or when serial data is output (DO). |  |
| CL | 5 | Clock | - Used as the synchronization clock when serial data is input to the LC72130 (DI), or when serial data is output (DO). |  |
| DI | 4 | Data input | - Inputs serial data sent from the controller to the LC72130. | A02600 |
| DO | 6 | Data output | - Outputs serial data sent from the LC72130 to the controller. <br> The content of the output data is determined by the serial data DOC0 to DOC2. |  |
| $V_{D D}$ | 16 | Power supply | - The LC72130 power supply ( $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V ) <br> - The power on reset circuit operates when power is first applied. | - |

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| Symbol | Pin No. | Type | Functions | Circuit configuration |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{S S}$ | 23 | Ground | - The LC72130 ground | - |
| $\overline{\mathrm{BO}}$ $\overline{\mathrm{BO} 2}$ $\overline{\mathrm{BO} 3}$ $\overline{\mathrm{BO} 4}$ $\overline{\mathrm{BO} 5}$ | $\begin{gathered} 7 \\ 8 \\ 9 \\ 10 \\ 2 \end{gathered}$ | Output port | - Dedicated output pins <br> - The output states are determined by BO 1 to BO 5 in the serial data. <br> Data: $0=$ open, 1 = low <br> - These pins go to the open state after the power on reset. <br> - An 8 Hz time base signal can be output from $\overline{\mathrm{BO} 1}$ when TBC in the serial data is set to 1 . <br> - Note that the ON impedance of the $\overline{\mathrm{BO} 1}$ pin is higher than that of the other pins ( $\overline{\mathrm{BO} 2}$ to $\overline{\mathrm{BO} 5})$. |  |
| $\frac{\overline{\mathrm{IO} 1}}{\mathrm{IO} 2}$ | $\begin{aligned} & 11 \\ & 13 \end{aligned}$ | I/O port | - Pins used for both input and output <br> - The input or output state is determined by bits IOC1 and IOC2 in the serial data. <br> Data: $0=$ input port, 1 = output port <br> - When specified for use as an input port: The input state is transmitted to the controller through the DO pin. <br> Input state: Low $\rightarrow$ data value $=0$ $\text { High } \rightarrow \text { data value }=1$ <br> - When specified for use as an output port: <br> The output state is determined by bits IO 1 and IO 2 in the serial data. <br> Data: $0=$ open, 1 = low <br> - These pins go to the input port state after the power ON reset. | A02602 |
| $\begin{aligned} & \text { PD1 } \\ & \text { PD2 } \end{aligned}$ | $\begin{aligned} & 19 \\ & 20 \end{aligned}$ | Charge pump output | - PLL charge pump output When the frequency generated by dividing the local oscillator frequency by N is higher than the reference frequency, a high level will be output from the PD pin. Similarly, when that frequency is lower, a low level will be output. The PD pin goes to the high impedance state when the frequencies agree. |  |
| AIN1 <br> AOUT1 <br> AIN2 <br> AOUT2 | $\begin{aligned} & 18 \\ & 17 \\ & 21 \\ & 22 \end{aligned}$ | LPF amplifier transistor | - The MOS transistor used for the PLL active low-pass filter. |  |
| IFIN | 12 | IF counter | - The input frequency range is from 0.4 to 12 MHz . <br> - The signal is supplied directly to the IF counter. <br> - The result from the IF counter MSB is output through the DO pin. <br> - There are four measurement periods: $4,8,32$, or 64 ms . | A02599 |

## Serial Data I/O Methods

The LC72130 uses Sanyo's audio LSI serial bus format, the CCB (computer control bus) format, for data I/O. This LSI adopts an 8-bit address version of the CCB format.


1. DI Control Data (Serial Data Input)

- IN1 Mode

- IN2 Mode



## 2. DI Control Data Functions



Continued from preceding page.

| No. | Control block/data | Functions | Related data |
| :---: | :---: | :---: | :---: |
| (6) | DO pin control data DOC0, DOC1, DOC2 | - Data that determines the DO pin output <br> The open state is selected after the power ON reset. <br> Note: 1. end-UC: Check for IF counter measurement completion <br> A02608 <br> (1) When end-UC is set and the IF counter is started (i.e., when CTE is changed from zero to one), The DO pin automatically goes to the open state. <br> (2) When the IF counter measurement completes, the DO pin goes low to indicate the measurement completion state. <br> (3) Depending on serial data I/O (CE: high) the DO pin goes to the open state. <br> 2. Goes to the open state if the I/O pin is specified to be an output port. <br> Caution: The state of the DO pin during a data input period (an IN1 or IN2 mode period with CE high) will be open, regardless of the state of the DO control data (DOC0 to DOC2). Also, the DO pin during a data output period (an OUT mode period with CE high) will output the contents of the internal DO serial data in synchronization with the CL pin signal, regardless of the state of the DO control data (DOC0 to DOC2). | ULO, UL1, CTE, IOC1, IOC2 |
| (7) | Unlock detection data ULO, UL1 | - Selects the phase error ( $\varnothing \mathrm{E}$ ) detection width for checking PLL lock. A phase error in excess of the specified detection width is seen as an unlocked state. <br> Note: In the unlocked state the DO pin goes low and the UL bit in the serial data becomes zero. | $\begin{aligned} & \text { DOC0, } \\ & \text { DOC1, } \\ & \text { DOC2 } \end{aligned}$ |
| (8) | Phase comparator control data DZ0, DZ1 | - Controls the phase comparator dead zone. <br> Dead zone widths: DZA < DZB < DZC < DZD |  |
| (9) | Clock time base TBC | Setting TBC to one causes an $8 \mathrm{~Hz}, 40 \%$ duty clock time base signal to be output from the $\overline{\mathrm{BO1}} \mathrm{pin}$. (BO1 data is invalid in this mode.) | BO1 |
| (10) | Charge pump control data DLC | - Forcibly controls the charge pump output. <br> Note: If deadlock occurs due to the VCO control voltage (Vtune) going to zero and the VCO oscillator stopping, deadlock can be cleared by forcing the charge pump output to low and setting Vtune to $\mathrm{V}_{\mathrm{CC}}$. (This is the deadlock clearing circuit.) |  |

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| No. | Control block/data | Functions | Related data |
| :---: | :--- | :--- | :--- |
| (11) | IF counter control data | - Note that if this value is set to zero the system enters input sensitivity degradation mode, <br> and the sensitivity is reduced to 10 to 30 mV rms. <br> * See the "IF Counter Operation" item for details. |  |
| (12) | LSI test data <br> TEST 0 to TEST2 | $\left.\begin{array}{l}\text { • LSI test data } \\ \text { TEST0 } \\ \text { TEST1 } \\ \text { TEST2 }\end{array}\right]$ These values must all be set to 0. |  |
| These test data are set to 0 automatically after the power ON reset. |  |  |  |$\quad$.

## 3. DO Output Data (Serial Data Output)

- OUT Mode



## 4. DO Output Data

| No. | Control block/data | Functions | Related data |
| :---: | :---: | :---: | :---: |
| (1) | I/O port data I2, I1 | - Latched from the pin states of the TOT and $\overline{\mathrm{O} 2} \mathrm{I} / \mathrm{O}$ ports. <br> - These values follow the pin states regardless of the input or output setting. <br> - Data is latched at the point where the circuit enters data output mode (OUT mode) <br> $11 \leftarrow \overline{\mathrm{O} 1}$ pin state $\backslash \mathrm{High}: 1$ <br> $\mathrm{I} 2 \leftarrow \overline{\mathrm{IO} 2}$ pin state JLow: 0 | $\begin{aligned} & \text { IOC1, } \\ & \text { IOC2, } \end{aligned}$ |
| (2) | PLL unlock data UL | - Latched from the state of the unlock detection circuit. UL $\leftarrow 0$ : Unlocked <br> $\mathrm{UL} \leftarrow 1$ : Locked or detection stopped mode | ULO, <br> UL1 |
| (3) | IF counter binary data C19 to C0 | - Latched from the value of the IF counter (20-bit binary counter). $\mathrm{C} 19 \leftarrow \mathrm{MSB}$ of the binary counter $\mathrm{C} 0 \leftarrow \mathrm{LSB}$ of the binary counter | CTE, GT0, GT1 |

5. Serial Data Input (IN1/IN2) $\mathrm{t}_{\mathrm{SU}}, \mathrm{t}_{\mathrm{HD}}, \mathrm{t}_{\mathrm{EL}}, \mathrm{t}_{\mathrm{ES}}, \mathrm{t}_{\mathrm{EH}} \geq 0.75 \mu \mathrm{~s}, \mathrm{t}_{\mathrm{LC}} \leq 0.75 \mu \mathrm{~s}$
(1) CL: Normal high

(2) CL: Normal low

6. Serial Data Output (OUT) $\mathrm{t}_{\mathrm{SU}}, \mathrm{t}_{\mathrm{HD}}, \mathrm{t}_{\mathrm{EL}}, \mathrm{t}_{\mathrm{ES}}, \mathrm{t}_{\mathrm{EH}} \geq 0.75 \mu \mathrm{~s}, \mathrm{t}_{\mathrm{DC}}, \mathrm{t}_{\mathrm{DH}} \leq 0.35 \mu \mathrm{~s}$ (1) CL: Normal high

(2) CL: Normal low


Note: Since the DO pin is an n-channel open drain pin, the time for the data to change ( $\mathrm{t}_{\mathrm{DC}}$ and $\mathrm{t}_{\mathrm{DH}}$ ) will differ depending on the value of the pull-up resistor and printed circuit board capacitance.
7. Serial Data Timing


| Parameter | Symbol | Pins | Conditions | min | typ | max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Data setup time | tsu | DI, CL |  | 0.75 |  |  | $\mu \mathrm{s}$ |
| Data hold time | $\mathrm{t}_{\mathrm{HD}}$ | DI, CL |  | 0.75 |  |  | $\mu \mathrm{s}$ |
| Clock low-level time | $\mathrm{t}_{\mathrm{CL}}$ | CL |  | 0.75 |  |  | $\mu \mathrm{s}$ |
| Clock high-level time | ${ }_{\text {t }}$ | CL |  | 0.75 |  |  | $\mu \mathrm{s}$ |
| CE wait time | $\mathrm{t}_{\mathrm{EL}}$ | CE, CL |  | 0.75 |  |  | $\mu \mathrm{s}$ |
| CE setup time | tes | CE, CL |  | 0.75 |  |  | $\mu \mathrm{s}$ |
| CE hold time | $\mathrm{t}_{\mathrm{EH}}$ | CE, CL |  | 0.75 |  |  | $\mu \mathrm{s}$ |
| Data latch change time | tLC |  |  |  |  | 0.75 | $\mu \mathrm{s}$ |
| Data output time | ${ }^{\text {D }}$ C | DO, CL | Differs depending on the value of the pull-up resistor and the printed circuit board capacitance. |  |  | 0.35 | $\mu \mathrm{s}$ |
|  | ${ }^{\text {D }}$ H | DO, CE |  |  |  |  |  |

## Programmable Divider



A026 16

|  | DVS | SNS | Input pin | Set divisor | Actual divisor: $N$ | Input frequency range $(\mathrm{MHz})$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A | 1 | $*$ | FMIN | 272 to 65535 | Twice the set value | 10 to 160 |
| B | 0 | 1 | AMIN | 272 to 65535 | The set value | 2 to 40 |
| C | 0 | 0 | AMIN | 4 to 4095 | The set value | 0.5 to 10 |

Note: * Don't care.

1. Programmable Divider Calculation Examples

- FM, 50 kHz steps (DVS = 1, SNS $=$ *, FMIN selected)

FM RF $=90.0 \mathrm{MHz}(\mathrm{IF}=+10.7 \mathrm{MHz})$
FM VCO $=100.7 \mathrm{MHz}$
PLL fref $=25 \mathrm{kHz}(\mathrm{R} 0$ to $\mathrm{R} 1=1, \mathrm{R} 2$ to $\mathrm{R} 3=0)$
100.7 MHz (FM VCO) $\div 25 \mathrm{kHz}($ fref $) \div 2$ (FMIN: divide-by-two prescaler) $=2014 \rightarrow 07 \mathrm{DE}(\mathrm{HEX})$


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- $\mathrm{SW}, 5 \mathrm{kHz}$ steps ( $\mathrm{DVS}=0, \mathrm{SNS}=1$, AMIN high speed side selected )

SW RF $=21.75 \mathrm{MHz}(\mathrm{IF}=+450 \mathrm{kHz})$
SW VCO $=22.20 \mathrm{MHz}$
PLL fref $=5 \mathrm{kHz}(\mathrm{R} 0=\mathrm{R} 2=0, \mathrm{R} 1=\mathrm{R} 3=1)$
$22.2 \mathrm{MHz}(\mathrm{SW} \mathrm{VCO}) \div 5 \mathrm{kHz}($ fref $)=4440 \rightarrow 1158(\mathrm{HEX})$


402818

- MW, 10 kHz steps (DVS $=0, \mathrm{SNS}=0$, AMIN low-speed side selected $)$

MW RF $=1000 \mathrm{kHz}(\mathrm{IF}=+450 \mathrm{kHz})$
MW VCO $=1450 \mathrm{kHz}$
PLL fref $=10 \mathrm{kHz}(\mathrm{R} 0$ to $\mathrm{R} 2=0, \mathrm{R} 3=1)$
$1450 \mathrm{kHz}(\mathrm{MW} \mathrm{VCO}) \div 10 \mathrm{kHz}($ fref $)=145 \rightarrow 091(\mathrm{HEX})$

| $\overbrace{}^{1} \overbrace{0}^{0}$ |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| * | * | * | * | 1 | - | - | - | 1 | 0 | 0 | 1 | o | - | - | 0 | - | - |  |  | 0 | 0 | - | 1 |
| 일 | $\bar{a}$ | ~ | \% | $\stackrel{\rightharpoonup}{\square}$ | a | 0 | a | \% | \% | $\stackrel{\circ}{\square}$ | a | $\frac{a}{a}$ | $\frac{m}{a}$ | $\frac{\square}{\square}$ | $\stackrel{10}{2}$ | $\begin{gathered} -\infty \\ \stackrel{e}{e} \\ \hline \end{gathered}$ | $\begin{aligned} & n \\ & 20 \\ & 0 \end{aligned}$ | $\stackrel{\leftarrow}{5}$ | $\stackrel{\text { a }}{\times}$ | \% | $\stackrel{\square}{\sim}$ | $\underset{\sim}{\sim}$ | $\stackrel{y}{x}$ |

## IF Counter

The LC72130 IF counter is a 20-bit binary counter. The result, i.e., the counter's msb, can be read serially from the DO pin.


| GT1 | GT0 | Measurement time |  |
| :---: | :---: | :---: | :---: |
|  |  | Measurement period (GT) (ms) | Wait time (twu) (ms) |
| 0 | 0 | 4 | 3 to 4 |
| 0 | 1 | 8 | 3 to 4 |
| 1 | 0 | 32 | 7 to 8 |
| 1 | 1 | 64 | 7 to 8 |

The IF frequency ( Fc ) is measured by determining how many pulses were input to an IF counter in a specified measurement period, GT.
$\mathrm{Fc}=\frac{\mathrm{C}}{\mathrm{GT}}$
$(\mathrm{C}=\mathrm{Fc} \times \mathrm{GT})$
C: Count value (number of pulses)

1. IF Counter Frequency Calculation Examples

- When the measurement period (GT) is 32 ms , the count (C) is 53980 hexadecimal (342400 decimal): IF frequency $(\mathrm{Fc})=342400 \div 32 \mathrm{~ms}=10.7 \mathrm{MHz}$


A0262 1

- When the measurement period (GT) is 8 ms , the count (C) is E10 hexadecimal (3600 decimal): IF frequency $(\mathrm{Fc})=3600 \div 8 \mathrm{~ms}=450 \mathrm{kHz}$

|  |  |  |  |  |  |  |  | $\bigcirc$ |  |  | E |  |  |  |  |  |  |  |  | 0 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | - | - | 0 | - | $\bigcirc$ |  | $\bigcirc$ | 1 |  | 1 | - | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | 1 | $\bigcirc$ | $\bigcirc$ |  | 0 |
| $\sim$ | $\square$ | $\stackrel{5}{5}$ | $\stackrel{\square}{0}$ | - | $\hat{0}$ | \% | 0 | - | $\stackrel{\square}{0}$ | $\stackrel{\square}{0}$ | - | \% | \% | ¢ | $\bigcirc$ | \% | : | \% | \% | - | \% |

2. IF Counter Operation


Prior to starting the IF counter, reset the IF counter in advance by setting CTE in the serial data to zero.
The IF counter is started by changing the value of CTE in the serial data from zero to one. The serial data is latched when the CE pin is dropped from high to low. The IF signal must be supplied to the IFIN pin in the period between the point the CE pin goes low and the end of the wait time at the latest. Next, the value of the IF counter at the end of the measurement period must be read out during the period that CTE is 1 . This is because the IF counter is reset when CTE is set to 0 .

Note: When operating the IF counter, the control microprocessor must check for the presence of the IF-IC SD (station detect signal) and, must turn on the IF buffer output and operate the counter only if the SD signal is present. Autosearch techniques that use only the IF counter are not recommended, since it is possible for IF buffer leakage output to cause incorrect stops at points where there is no station.

## IFIN minimum input sensitivity standard

| IFS | $0.4 \leq \mathrm{f}<0.5$ | $0.5 \leq \mathrm{f}<8$ | $8 \leq \mathrm{f} \leq 12$ |
| :---: | :---: | :---: | :---: |
| 1: Normal mode | 40 mVrms <br> $(0.1$ to 3 mVrms$)$ | 40 mVrms | 40 mVrms <br> (1 to 10 mVrms$)$ |
| 0: Degradation mode | 70 mVrms <br> (10 to 15 mVrms$)$ | 70 mVrms | 70 mVrms <br> $(30$ to 40 mVrms$)$ |

[^0]
## Unlock Detection Timing

1. Unlock Detection Determination Timing

Unlocked state detection is performed in the reference frequency (fref) period (interval). Therefore, in principle, this determination must be performed over a period no less than the reference frequency period. However, directly following a change to the (frequency) divisor N , that determination must be performed after at least two reference frequency periods have passed.


Figure 1 Unlocked State Detection Timing
For example, if fref is 1 kHz , i.e., the period is 1 ms , after the divisor N is changed, unlocked state determination must be performed after waiting 2 ms .


Figure 2 Circuit Structure
2. Unlock Determination Software Integration Method


Figure 3
3. Unlocked State Data Output Using Serial Data Output

In the LC72130, once an unlocked state occurs, the unlocked state serial data (UL) will not be reset until a data input (or output) operation is performed. At the data output 1 point in Figure 3, although the VCO frequency has stabilized (locked), since no data output has been performed since the divisor N was changed the unlocked state data remains in the unlocked state. As a result, even though the frequency has stabilized (locked), the system remains (from the standpoint of the data) in the unlocked state.

Therefore, the unlocked state data acquired at data output 1 , which occurs immediately after the divisor N was changed, should be treated as a dummy data output and ignored. The second data output (data output 2 ) and following outputs are valid data.


Wait for at least two reference frequency periods.

Valid data can be output at intervals of one reference frequency period or longer.

Note: Locking state determination is more reliable if it is based on reading valid output data several times.

## Locked State Determination Flowchart

4. Directly Outputting Unlocked State Data from the DO Pin (Set by the DO pin control data)

Since the locking state ( high $=$ locked, low $=$ unlocked ) is output directly from the DO pin, the dummy data processing described in section 3 above is not required. After changing the divisor N , the locking state can be checked after waiting at least two reference frequency periods.

## Clock Time Base Usage Notes

The pull-up resistor used on the clock time base output pin $(\overline{\mathrm{BO} 1})$ should be at least $100 \mathrm{k} \Omega$. Also, to prevent chattering we recommend using a Schmitt input at the controller (microprocessor) that receives this signal.

This is to prevent degrading the VCO C/N characteristics when a loop filter is formed using the built-in low-pass filter transistor. Since the clock time base output pin and the low-pass filter have a common ground internal to the IC, it is necessary to minimize the time base output pin current fluctuations and to suppress their influence on the low-pass filter.


## Other Items

1. Notes on the Phase Comparator Dead Zone

| DZ1 | DZ0 | Dead zone mode | Charge pump | Dead zone |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | DZA | ON/ON | --0 s |
| 0 | 1 | DZB | ON/ON | -0 s |
| 1 | 0 | DZC | OFF/OFF | +0 s |
| 1 | 1 | DZD | OFF/OFF | ++0 s |

Since correction pulses are output from the charge pump even if the PLL is locked when the charge pump is in the ON/ON state, the loop can easily become unstable. This point requires special care when designing application circuits.

The following problems may occur in the ON/ON state.

- Side band generation due to reference frequency leakage
- Side band generation due to both the correction pulse envelope and low frequency leakage

Schemes in which a dead zone is present (OFF/OFF) have good loop stability, but have the problem that acquiring a high $\mathrm{C} / \mathrm{N}$ ratio can be difficult. On the other hand, although it is easy to acquire a high $\mathrm{C} / \mathrm{N}$ ratio with schemes in which there is no dead zone, it is difficult to achieve high loop stability. Therefore, it can be effective to select DZA or DZB, which have no dead zone, in applications which require an FM S/R ratio in excess of 90 to 100 dB , or in which an increased AM stereo pilot margin is desired. On the other hand, we recommend selecting DZC or DZD, which provide a dead zone, for applications which do not require such a high FM signal-to-noise ratio and in which either AM stereo is not used or an adequate AM stereo pilot margin can be achieved.

## Dead Zone

The phase comparator compares fp to a reference frequency (fr) as shown in Figure 4. Although the characteristics of this circuit (see Figure 5) are such that the output voltage is proportional to the phase difference $\varnothing$ (line A), a region (the dead zone) in which it is not possible to compare small phase differences occurs in actual ICs due to internal circuit delays and other factors (line B). A dead zone as small as possible is desirable for products that must provide a high $\mathrm{S} / \mathrm{N}$ ratio.
However, since a larger dead zone makes this circuit easier to use, a larger dead zone is appropriate for popularlypriced products. This is because it is possible for RF signals to leak from the mixer to the VCO and modulate the VCO in popularly-priced products in the presence of strong RF inputs. When the dead zone is narrow, the circuit outputs correction pulses and this output can further modulate the VCO and generate beat frequencies with the RF signal.


Figure 4
402930


Figure 5

A02931
2. Notes on the FMIN, AMIN, and IFIN Pins

Coupling capacitors must be placed as close as possible to their respective pin. A capacitance of about 100 pF is desirable. In particular, if a capacitance of 1000 pF or over is used for the IF pin, the time to reach the bias level will increase and incorrect counting may occur due to the relationship with the wait time.
3. Notes on IF Counting $\rightarrow$ SD must be used in conjunction with the IF counting time

When using IF counting, always implement IF counting by having the microprocessor determine the presence of the IF-IC SD (station detect) signal and turn on the IF counter buffer only if the SD signal is present. Schemes in which auto-searches are performed with only IF counting are not recommended, since they can stop at points where there is no signal due to leakage output from the IF counter buffer.
4. DO Pin Usage Techniques

In addition to data output mode times, the DO pin can also be used to check for IF counter count completion and for unlock detection output. Also, an input pin state can be output unchanged through the DO pin and input to the controller.
5. Power Supply Pins

A capacitor of at least 2000 pF must be inserted between the power supply $\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{SS}}$ pins for noise exclusion. This capacitor must be placed as close as possible to the $\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{SS}}$ pins.

Pin States After the Power ON Reset


## Sample Application System



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[^0]:    ( ): Actual values (reference data)

